## Chapter 9

## **Assignment 9 Solutions**

- **6.5** Consider the avalanche device of the preceding problem.
- a. Plot the S/N vs. M for values of M ranging from 1 to 100.
- b. Find the approximate value of  $M_{\rm opt}$  and S/N at  $M=M_{\rm opt}$  from your graph.
- c. Calculate the values of  $M_{\rm opt}$  and S/N at  $M=M_{\rm opt}$  from formulas and compare with your graphical results of the previous question.

Solution: a) (Note: We will use  $F(M) = M^x = M^{0.5}$  as in the previous problem.) The signal-to-noise ratio is given by

$$\frac{S}{N} = \frac{(m\mathcal{R}_0 P_0 M)^2 / 2}{2q\mathcal{R}_0 P_0 B M^{2.5} + 2q i_D B M^{2.5} + 2q I_s B + (4kTB/R_L)}.$$
(9.1)

We know all of the values but M,

$$\langle i_s^2 \rangle = \frac{(0.85)(0.583)(1 \times 10^{-8})(M))^2}{2} = 1.228 \times 10^{-17} M^2 \text{ A}^2,$$
 (9.2)

$$\langle i_N^2 \rangle_1 = 2q \mathcal{R}_0 P_0 B M^{2.5}$$
 (9.3)  
=  $(2)(1.6 \times 10^{-19})(0.583)(1 \times 10^{-8})(1 \times 10^4) M^{2.5}$   
=  $1.868 \times 10^{-23} M^{2.5} A^2$ .

$$\langle i_N^2 \rangle_2 = 2qI_D B M^{2.5}$$
 (9.4)  
=  $(2)(1.6 \times 10^{-19})(1 \times 10^{-9})(1 \times 10^4) M^{2.5}$   
=  $3.204 \times 10^{-24} M^{2.5} A^2$ ,

$$\langle i_N^2 \rangle_3 = 2qI_sB = (2)(1.6 \times 10^{-19})(1 \times 10^{-9})(1 \times 10^4)$$
  
=  $3.204 \times 10^{-24} \text{ A}^2$ , (9.5)

and

$$\langle i_N^2 \rangle_4 = \frac{4kTB}{R_L} = \frac{(4)(1.38 \times 10^{-23})(300)(1 \times 10^4)}{1 \times 10^4}$$
  
= 1.657 × 10<sup>-20</sup> A<sup>2</sup>. (9.6)

So we can write the SNR as

$$\frac{S}{N} = \frac{1.228 \times 10^{-17} M^2}{1.868 \times 10^{-23} M^{2.5} + 3.204 \times 10^{-24} M^{2.5} + 3.204 \times 10^{-24} + 1.657 \times 10^{-20}}$$

$$= \frac{1.228 \times 10^{-17} M^2}{(2.19 \times 10^{-23}) M^{2.5} + 1.656 \times 10^{-20}}.$$

This function is plotted in Fig. 9.1.

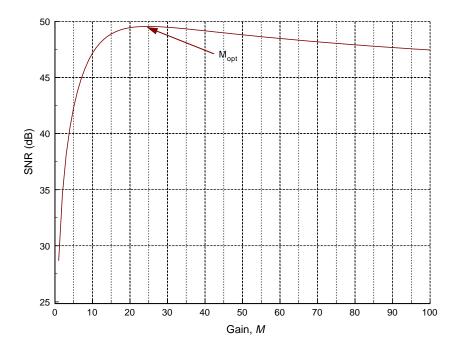


Figure 9.1: Plot of S/N vs. M for Prob. 6.4a.

b) From the plot we see that  $M_{\rm opt} \approx 25$  and that the maximum signal-to-noise ratio is about 49.5 dB.

The *computed* value of  $M_{\rm opt}$  is found from

$$M_{\text{opt}}^{x+2} \approx \frac{2qI_s + (4kT/R_L)}{xq(\mathcal{R}_0 P_0 + I_D)}$$
 (9.7)

$$M_{\text{opt}}^{2.5} \approx \frac{(2)(1.6 \times 10^{-19})(1 \times 10^{-9}) + \frac{(4)(1.38 \times 10^{-23})(300)}{1 \times 10^4}}{(0.5)(1.6 \times 10^{-19})((0.583)(1 \times 10^{-8}) + 1 \times 10^{-9})}$$
  
 $\approx 3.03 \times 10^3$   
 $M_{\text{opt}} \approx (3.03 \times 10^3)^{1/2.5} = 24.7$ .

The signal-to-noise ratio at the optimum value of M is

$$\frac{S}{N} = \frac{1.228 \times 10^{-17} M_{\text{opt}}^2}{(2.19 \times 10^{-23}) M_{\text{opt}}^{2.5} + 1.656 \times 10^{-20}} 
= \frac{1.228 \times 10^{-17} (24.7)^2}{(2.19 \times 10^{-23}) (24.7)^{2.5} + 1.656 \times 10^{-20}} 
= 9.03 \times 10^4 \Rightarrow 49.55 \text{ dB}$$
(9.8)

These calculated results compare favorably with the graphical results.

 ${f 6.6.}$  Consider a silicon avalanche photodiode with parameters as given below operating in a link with no intersymbol interference present.

$$\begin{split} F(M) &= M^{0.4} \\ \text{Responsivity (at } M=1) = 0.3 \text{ A/W} \\ \text{Surface dark current} &= 1 \ \mu\text{A} \\ \text{Temperature} &= 300 \text{ K} \\ R_L &= 1 \text{ k}\Omega \\ \text{Bulk dark current} &= 1 \text{ nA} \\ \text{Bandwidth of receiver} &= 10 \text{ MHz} \end{split}$$

- a. Calculate the dc optical power that must be incident on the detector to make the optimum gain of this amplifier have a value of 80.
- b. With a value of gain of 80 calculate the ratio (in dB) of the mean-square noise current due to the shot noise caused by the bulk dark current to the mean-square noise current due to the thermal noise.

Solution: We are given that 
$$F(M)=M^{0.4},~\mathcal{R}_0=0.3,~I_s=1\times 10^{-6}~\mathrm{A},~T=300\mathrm{K},~R_L=1\times 10^3,~I_D=1\times 10^{-9},~\mathrm{and}~B=1\times 10^6.$$

a. We are given that  $M_{\mathrm{opt}}=80$  and find the power P from

$$M_{\text{opt}}^{x+2} \approx \frac{2qI_s + (4kT/R_L)}{xq(\mathcal{R}_0 P_0 + I_D)}$$
 (9.9)

$$(80)^{2.4} \approx \frac{(2)(1.6 \times 10^{-19})(1 \times 10^{-6}) + \frac{(4)(1.38 \times 10^{-23})(300)}{1 \times 10^{3}}}{(0.4)(1.6 \times 10^{-19})(0.3P + 1 \times 10^{-9})}, \qquad (9.10)$$

SO

$$0.3P + 1 \times 10^{-9} = \frac{1.692 \times 10^{-23}}{(0.4)(1.6 \times 10^{-19})(80)^{2.4}} = 7.159 \times 10^{-9}$$

$$0.3P = 6.159 \times 10^{-9}$$

$$P = 2.05 \times 10^{-8} \text{ W} = 20.5 \text{ nW}.$$

$$(9.11)$$

b. The ratio of the shot noise due to the dark current  $(< i_N^2>_1)$  to the noise due to thermal noise  $(< i_N^2>_2)$  is

$$\frac{\langle i_N^2 \rangle_1}{\langle i_N^2 \rangle_2} = \frac{2qI_D B M^2 F(M)}{4kTB/R_L}$$

$$= \frac{(2)(1.6 \times 10^{-19})(1 \times 10^{-9})(80)^{2.4}}{[(4)(1.38 \times 10^{-23})(300)/(1 \times 10^3)]}$$

$$= 0.710 \Rightarrow -1.466 \text{ dB}.$$
(9.12)

- **6.7.** Consider a silicon photodiode operating at 850 nm ( $\alpha = 10^3 \text{ cm}^{-1}$ ).
- a. Calculate the area of the device if the capacitance is to be kept equal or less than 2 pF. The relative permittivity of silicon is 11.7.
- b. Calculate the maximum bandwidth of the detector when operating into 50  $\Omega$  load.

Solution: a) We estimate the width w of the depletion region as

$$w = \frac{2}{\alpha} = \frac{2}{1 \times 10^5} = 2 \times 10^{-5} \text{ m}.$$
 (9.13)

The capacitance is required to obey  $C \le 2 \times 10^{-12}$ , so

$$C_{\text{max}} = 2 \times 10^{-12} \tag{9.14}$$

$$\frac{\epsilon_r \epsilon_0 A_{\text{max}}}{w} = 2 \times 10^{-12} \tag{9.15}$$

$$A_{\text{max}} = \frac{(2 \times 10^{-12})w}{\epsilon_r \epsilon_0} = \frac{(2 \times 10^{-12})(2 \times 10^{-5})}{(11.7)(8.85 \times 10^{-12})} = 3.86 \times 10^{-3} \text{ m}^2.$$
 (9.16)

We can find the diameter of a circular detector from

$$A_{\text{max}} = \frac{\pi d_{\text{max}}^2}{4} \tag{9.17}$$

$$d_{\text{max}} = \sqrt{\frac{4A_{\text{max}}}{\pi}} = \sqrt{\frac{(4)(3.86 \times 10^{-3})}{\pi}} = 701 \times 10^{-4} \text{m} = 0.701 \ \mu\text{m}. \tag{9.18}$$

The detector is very small, indeed!!

b) The maximum frequency that this detector can respond to when used with a 50  $\Omega$  load is

$$f_{\text{max}} = \frac{1}{2\pi R_L C} = \frac{1}{2\pi (50)(2 \times 10^{-12})} = 1.591 \times 10^9 \text{ Hz} = 1.591 \text{ GHz}.$$
 (9.19)

- 6.8. Consider a pin diode with the following properties at 920 nm—
  - Responsivity = 0.5 A/W
  - Dark current = 1.0 nA
  - Surface dark current is negligible
  - Operating temperature = 300 K.

This diode is irradiated with a constant 80 nW of optical power (at 920 nm). Find the signal-to-noise ratio (in dB) of the detector if it is operated into an equivalent load of  $10~\text{K}\Omega$  and a (noisefree) preamp with a bandwidth of 1~MHz.

Solution: The mean-square signal is found from

$$\langle i_s^2 \rangle = (\mathcal{R}_0 P)^2 = ((0.5)(8 \times 10^{-8}))^2 = 1.6 \times 10^{-15} \text{ A}^2.$$
 (9.20)

The mean-square noise current is found as

$$\langle i_N^2 \rangle = 2q \left( \mathcal{R}_0 P + I_D \right) B + 2q I_{\text{surface}} B + \frac{4kTB}{R_L}$$

$$= 2q \left( (0.5)(8 \times 10^8) + 1 \times 10^{-9} \right) (1 \times 10^6) + 0$$

$$+ \frac{4(1.38 \times 10^{-23})(300)(1 \times 10^6)}{1 \times 10^4}$$

$$= 1.669 \times 10^{-18} \text{ A}^2.$$

$$(9.21)$$

The signal-to-noise ratio is

$$\frac{S}{N} = \frac{1.6 \times 10^{-15}}{1.669 \times 10^{-18}} = 959 \Rightarrow 19.8 \text{ dB}. \tag{9.22}$$

6.9. Show that the relation between the Q-parameter and S/N is

$$Q = \frac{1}{2} \sqrt{\frac{S}{N}}. ag{9.23}$$

Solution: Equation 5.171 can be written as

BER = 
$$0.5 \operatorname{erf}\left(\frac{Q}{\sqrt{2}}\right)$$
. (9.24)

Equation 5.152 is

BER = 
$$0.5 \operatorname{erf}\left(\frac{\sqrt{\overline{SNR}}}{2\sqrt{2}}\right)$$
, (9.25)

so,

$$Q = \frac{\sqrt{\text{SNR}}}{2} = \frac{1}{2} \sqrt{\frac{S}{N}}.$$
 (9.26)